

# 1MHZ High Voltage Bipolar Opamp

#### **Features**

• Single-Supply Operation from +3V ~ +36V

Dual-Supply Operation from ±1.5V ~ ±18V

• Gain-Bandwidth Product: 1MHz (Typ)

Low Input Bias Current: 20nA (Typ)

Low Offset Voltage: 5mV (Max)

• Quiescent Current: 500µA per Amplifier (Typ)

Input Common Mode Voltage Range Includes

Ground

- Large Outpu Voltage Swing:0V to Vcc-1.5V
- Operating Temperature: -25°C ~ +85°C
- Small Package:

LM321H Available in SOT23-5 Package

LM358H Available in SOP-8 and MSOP-8 Packages

LM324H Available in SOP-14 Package

## **General Description**

The LM321H/358H/324H family have a high gain-bandwidth product of 1MHz, a slew rate of 0.4V/µs, and a quiescent current of 500µA/amplifier at 5V. The LM321H/358H/324H family is designed to provide optimal performance in low voltage and low noise systems. The maximum input offset voltage is 5mV for LM321H/358H/324H family. The operating range is from 3V to 36V. The LM321H single is available in Green SOT-23-5 package. The LM358H Dual is available in Green SOP-8 and MSOP-8 packages. The LM324H Quad is available in Green SOP-14 package.

## **Applications**

- Walkie-Talkie
- Battery Management Solution
- Transducer Amplifiers
- Summing Amplifiers

- Multivibrators
- Oscillators
- Switcching Telephone
- Portable Systems

# **Pin Configuration**

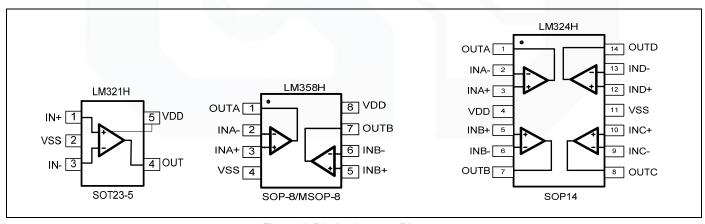


Figure 1. Pin Assignment Diagram



March 2020-REV V3



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# **Absolute Maximum Ratings**

Condition	Symbol	Max
Power Supply Voltage	Vcc	$\pm 20$ V or 40V
Differential input voltage	V <sub>I(DIFF)</sub>	40V
Input Voltage	Vı	-0.3V~40V
Operating Temperature Range	Topr	-25°C ~+85°C
Storage Temperature Range	Tstg	-65°C ~+150°C

**Note:** Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

# Package/Ordering Information

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
LM321H	Single	LM321H-TR	SOT23-5	Tape and Reel,3000	LM321
LM050II Decil	LM358H-SR	SOP-8	Tape and Reel,4000	LM358	
LM358H	Dual	LM358H-MR	MSOP-8	Tape and Reel,3000	LM358
LM324H	Quad	LM324H-SR	SOP-14	Tape and Reel,2500	LM324



# **Electrical Characteristics**

(At  $V_S = +15V$ ,  $T_A=25$ °C, unless otherwise noted.)

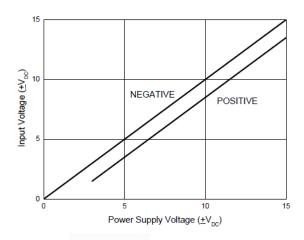
			LM321H/358H/324H			
PARAMETER	SYMBOL	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE		
			+25℃	+25℃	UNITS	MIN/MAX
INPUT CHARACTERISTICS					•	
Input Offset Voltage	Vos	$V_{CM} = V_S/2$	0.4	5	mV	MAX
Input Bias Current	I <sub>B</sub>		20		nA	TYP
Input Offset Current	Ios		5		nA	TYP
Common-Mode Voltage Range	V <sub>CM</sub>	V <sub>S</sub> = 5.5V	-0.1 to +4		V	TYP
Common-Mode Rejection Ratio	CMRR	V <sub>CM</sub> = 0V to Vs-1.5V	90	70	dB	MIN
Open-Loop Voltage Gain	A <sub>OL</sub>	$R_L = 5k\Omega$ , $V_O = 1V$ to 11V	100	85	dB	MIN
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$		7		μV/°C	TYP
OUTPUT CHARACTERISTICS						
Output Voltage Swing from Rail	V <sub>OH</sub>	$R_L = 2k\Omega$	11		٧	MIN
	V <sub>OL</sub>	$R_L = 2k\Omega$	5	20	mV	MAX
	V <sub>OH</sub>	R <sub>L</sub> = 10kΩ	12	13	V	MIN
	V <sub>OL</sub>	R <sub>L</sub> = 10kΩ	5	20	mV	MAX
	I <sub>SOURCE</sub>	D 400 to 1/ /0	40	60	- mA	MAX
Output Current	I <sub>SINK</sub>	$R_L = 10\Omega$ to $V_S/2$	40	60		
POWER SUPPLY						
0 " 1/1" D				3	٧	MIN
Operating Voltage Range				30	V	MAX
Power Supply Rejection Ratio	PSRR	$V_S = +5V \text{ to } +36V, V_{CM} = +0.5V$	100	70	dB	MIN
Quiescent Current / Amplifier	ΙQ	V <sub>S</sub> = 36V, RL=∞	0.5	2.5	mA	MAX
DYNAMIC PERFORMANCE	l		1	l		
Gain-Bandwidth Product	GBP		1		MHz	TYP
Slew Rate	SR	G = +1, 2V Output Step	0.4		V/µs	TYP



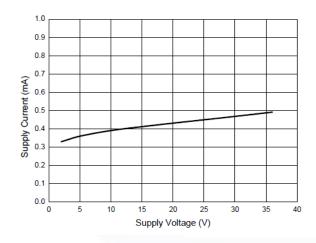


# **Typical Performance characteristics**

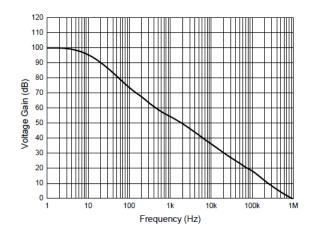
### Input Voltage Range



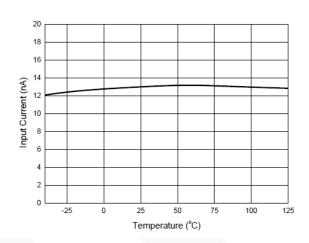
### **Supply Current**



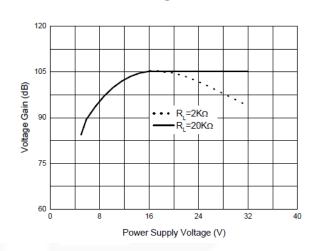
**Open Loop Frequency Response** 



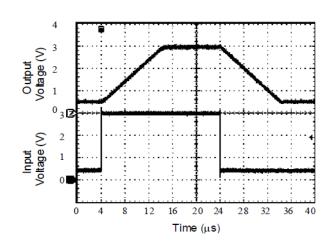
### **Input Current**



### Voltage Gain



Voltage Follower Pulse Response

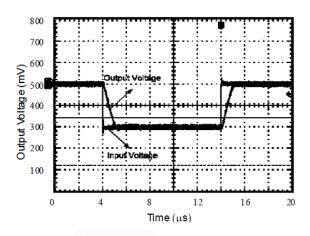




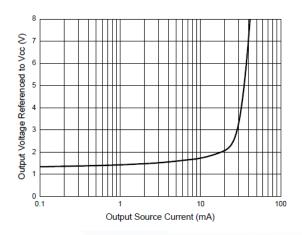


# **Typical Performance characteristics**

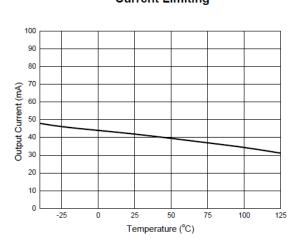
### Voltage Follower Pulse Response (Small Signal)



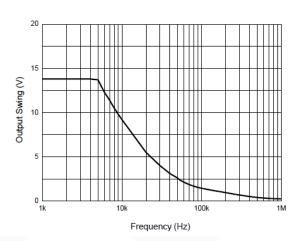
#### **Output Characteristics: Current Sourcing**



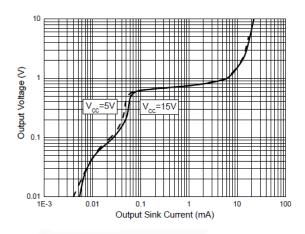
# **Current Limiting**



#### Large Signal Frequency Response



**Output Characteristics: Current Sinking** 





# **Application Note**

#### **Size**

LM321H/358H/324H family series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the LM321H/358H/324H family packages save space on printed circuit boards and enable the design of smaller electronic products.

#### **Power Supply Bypassing and Board Layout**

LM321H/358H/324H family series operates from a single 3V to 36V supply or dual  $\pm 1.5$ V to  $\pm 18$ V supplies. For best performance, a  $0.1\mu F$  ceramic capacitor should be placed close to the  $V_{DD}$  pin in single supply operation. For dual supply operation, both  $V_{DD}$  and  $V_{SS}$  supplies should be bypassed to ground with separate  $0.1\mu F$  ceramic capacitors.

### **Low Supply Current**

The low supply current (typical 500µA per channel) of LM321H/358H/324H family will help to maximize battery life.

#### **Operating Voltage**

LM321H/358H/324H family operates under wide input supply voltage (3V to 36V). In addition, all temperature specifications apply from -25 °C to +85 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime.

#### **Capacitive Load Tolerance**

The LM321H/358H/324H family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create apole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

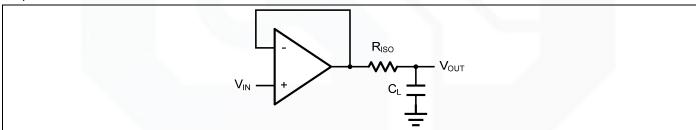


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor



The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the  $V_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

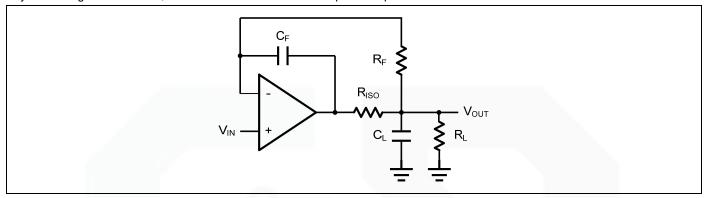


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy







# **Typical Application Circuits**

## Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using LM321H/358H/324H family.

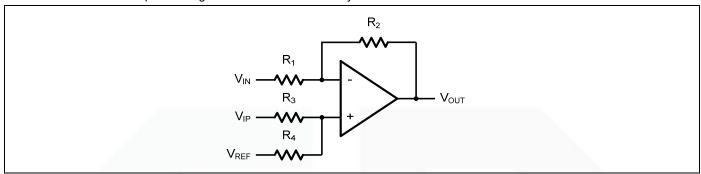


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R<sub>1</sub>=R<sub>3</sub> and R<sub>2</sub>=R<sub>4</sub>), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

### **Low Pass Active Filter**

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C=1/(2\pi R_3C_1)$ .

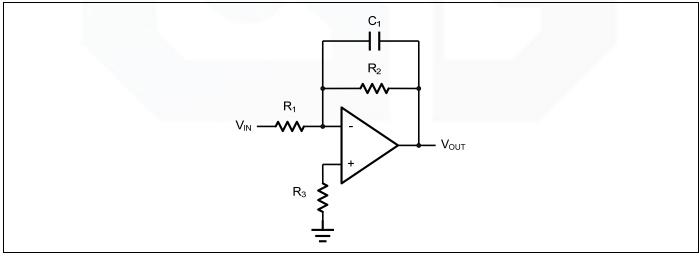


Figure 5. Low Pass Active Filter





# **Instrumentation Amplifier**

The triple LM321H/358H/324H family can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R2/R1. The two differential voltage followers assure the high input impedance of the amplifier.

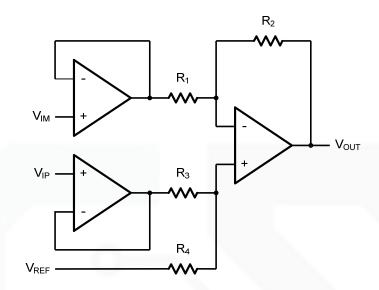
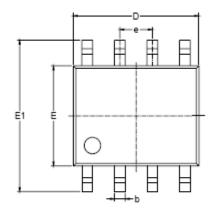


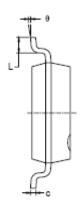
Figure 6. Instrument Amplifier

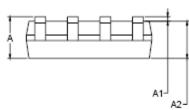


# **Package Information**

# SOP-8



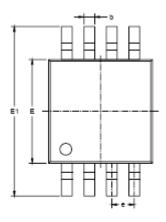




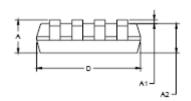
Symbol		nsions imeters	Dimensions In Inches		
	MIN	MAX	MIN	MAX	
A	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
с	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
e	1.27 BSC		0.050	BSC	
L	0.400	1.270	0.016	0.050	
е	0°	8°	0°	8°	



# MSOP-8



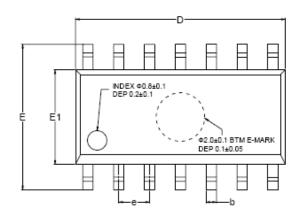


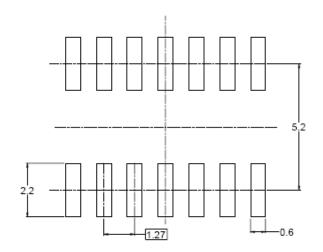


Symbol	Dimen In Milli		Dimensions In Inches		
-	MIN	MAX	MIN	MAX	
Α	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.006	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.650 BSC		0.026 BSC		
L	0.400	0.800	0.016	0.031	
θ	0°	6°	0°	6°	



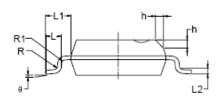
### SOP-14





RECOMMENDED LAND PATTERN (Unit: mm)

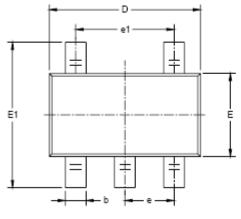


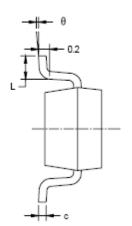


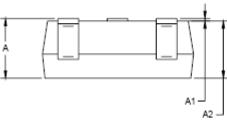
Symbol	Dimensions In Millimeters			Dimensions In Inches		
	MIN	MOD	MAX	MIN	MOD	MAX
Α	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.004		0.010
A2	1.25		1.65	0.049		0.065
A3	0.55		0.75	0.022		0.030
b	0.36		0.49	0.014		0.019
D	8.53		8.73	0.336		0.344
E	5.80		6.20	0.228		0.244
E1	3.80		4.00	0.150		0.157
е		1.27 BSC		0.050 BSC		
L	0.45		0.80	0.018		0.032
L1	1.04 REF			0.040 REF		
L2	0.25 BSC			0.01 BSC		
R	0.07			0.003		
R1	0.07			0.003		
h	0.30		0.50	0.012		0.020
θ	0°		8°	0°		8°



# SOT23-5







Symbol		sions imeters	Dimensions In Inches		
-,	MIN	MAX	MIN	MAX	
Α	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
E	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
e	0.950	BSC	0.037 BSC		
e1	1.900	1.900 BSC		BSC	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	